

Second-generation Micro-Spec: Spectrometer design for the Experiment for Cryogenic Large-Aperture Intensity Mapping



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Introduction

Micro-Spec (μ -Spec) is a direct-detection spectrometer which integrates all the components of a diffraction-grating spectrometer onto a $\sim 10\text{-cm}^2$, $<200\text{-g}$ chip through the use of **superconducting microstrip transmission lines on a single-crystal silicon substrate**.

The second generation of μ -Spec was designed and is currently being built to operate with a spectral resolution of 512 over a $714\text{-}555\text{ }\mu\text{m}$ ($420\text{-}540\text{ GHz}$) wavelength range, a band of interest for NASA's EXperiment for Cryogenic Large-Aperture Intensity Mapping (EXCLAIM).

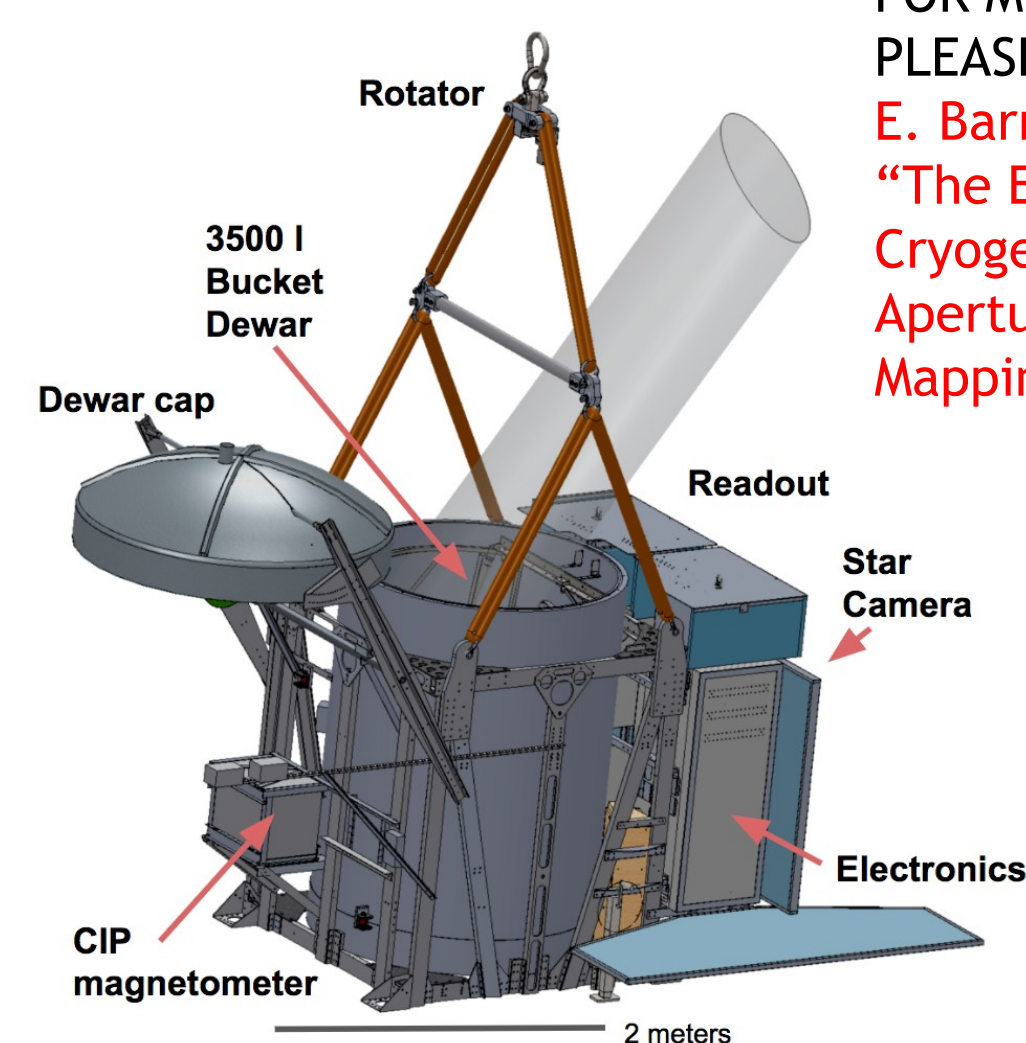
The EXCLAIM mission

EXCLAIM is a balloon-borne telescope that will map the emission of CO/[CII] lines in redshift windows of $0 < z < 3.5$. These lines are key tracers of the gas phases in the interstellar medium involved in star-formation processes.

EXCLAIM will adopt an approach called **Intensity Mapping (IM)**:

- It will measure the statistics of brightness fluctuations of redshifted, cumulative line emissions instead of detecting individual galaxies.
- This enables a blind, complete census of emitting gas in cross-correlation with a rich spectroscopic galaxy catalog such as the Baryon Oscillation Spectroscopic Survey (BOSS) [1].

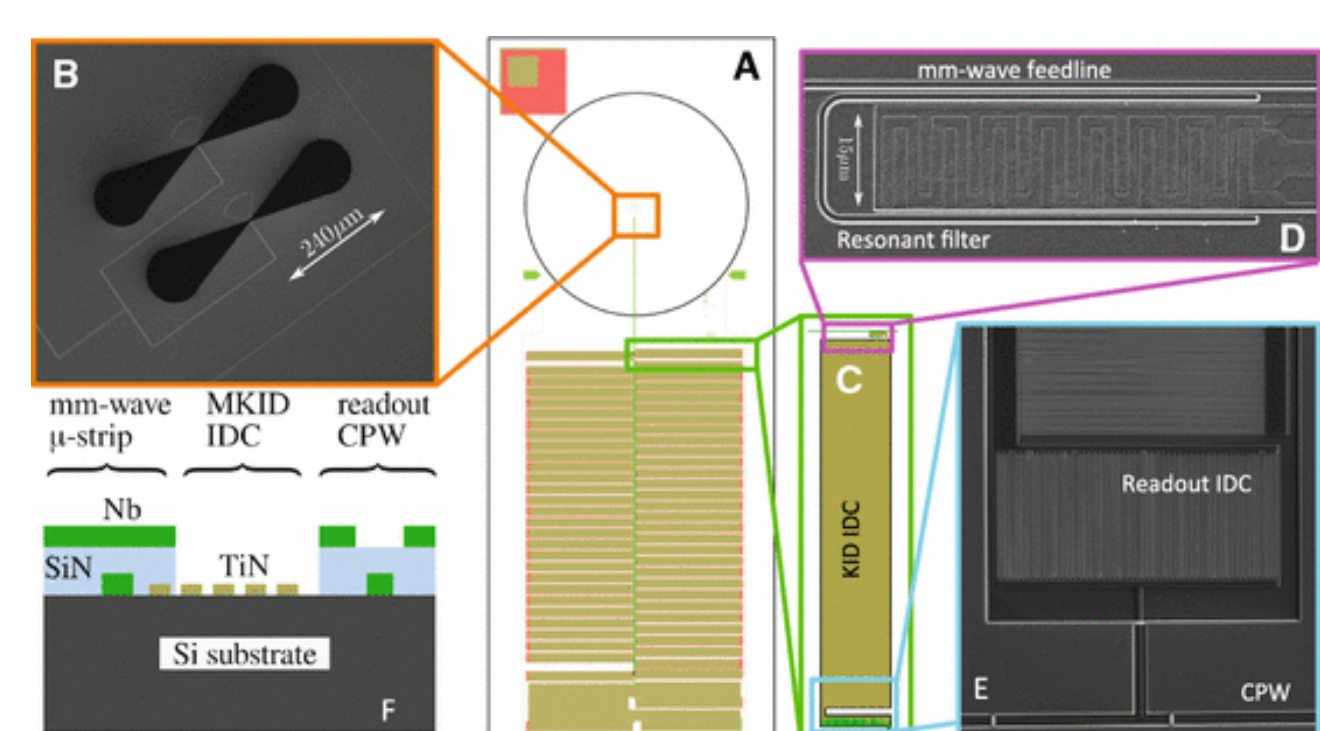
FOR MORE DETAILS, PLEASE SEE POSTER: E. Barrentine et al., "The Experiment for Cryogenic Large-Aperture Intensity Mapping (EXCLAIM)"



Literature review

μ -Spec differs from similar technologies.

- In a Rowland spectrometer, the required phase retardation is generated by reflection from the grating grooves [2].
- In Z-Spec, propagation occurs in parallel-plate waveguides [3], which remains a bulky option (55-cm length scale and 4-kg mass).
- Bootlace lenses are a 1-dimensional analog of Z-Spec [4], which μ -Spec builds on for submillimeter wave applications.
- Narrow-band filter-bank spectrometers do not rely on optical interference as in grating or Fabry-Perot spectrometers. Some examples are: SuperSpec [5, figure below], the Delft SRON High-redshift Mapper (DESHIMA) [6], the CAMbridge Emission Line Surveyor (CAMELS) [7], and similar alternatives made in rectangular waveguides (e.g., W-Spec [8]).



Instrument layout

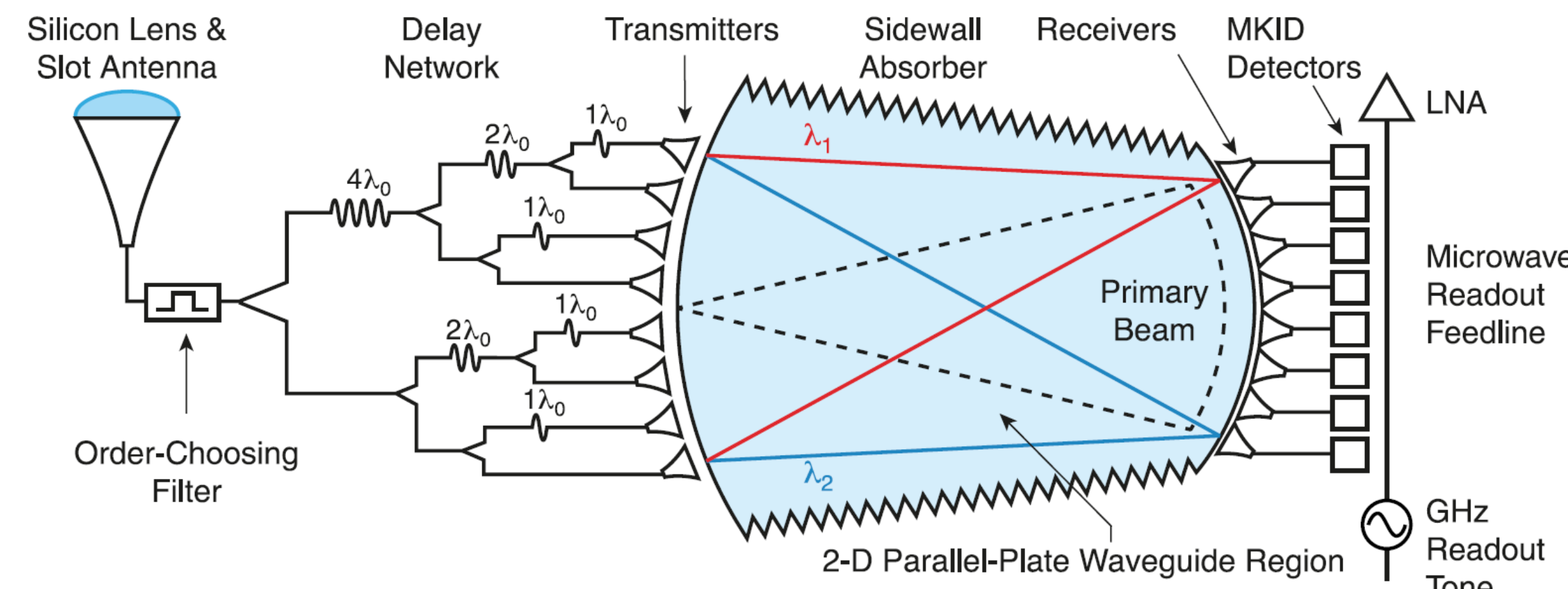


Figure 1. Layout of the μ -Spec module. The light is coupled into the instrument via a lens and a broadband antenna. An order-choosing filter selects the design order before the light is transmitted through a low-loss superconducting transmission line to a phase delay network (Fig. 7). The spectrum enters the planar diffractive region through an array of feed horn structures (transmitters), which concentrate the power along the focal surface as a function of wavelength. The receivers terminate in microwave kinetic inductance detectors (MKIDs) for readout. Sidewall absorbers terminate any power emitted into large angles or reflected from the receivers.

Building on demonstrated technology

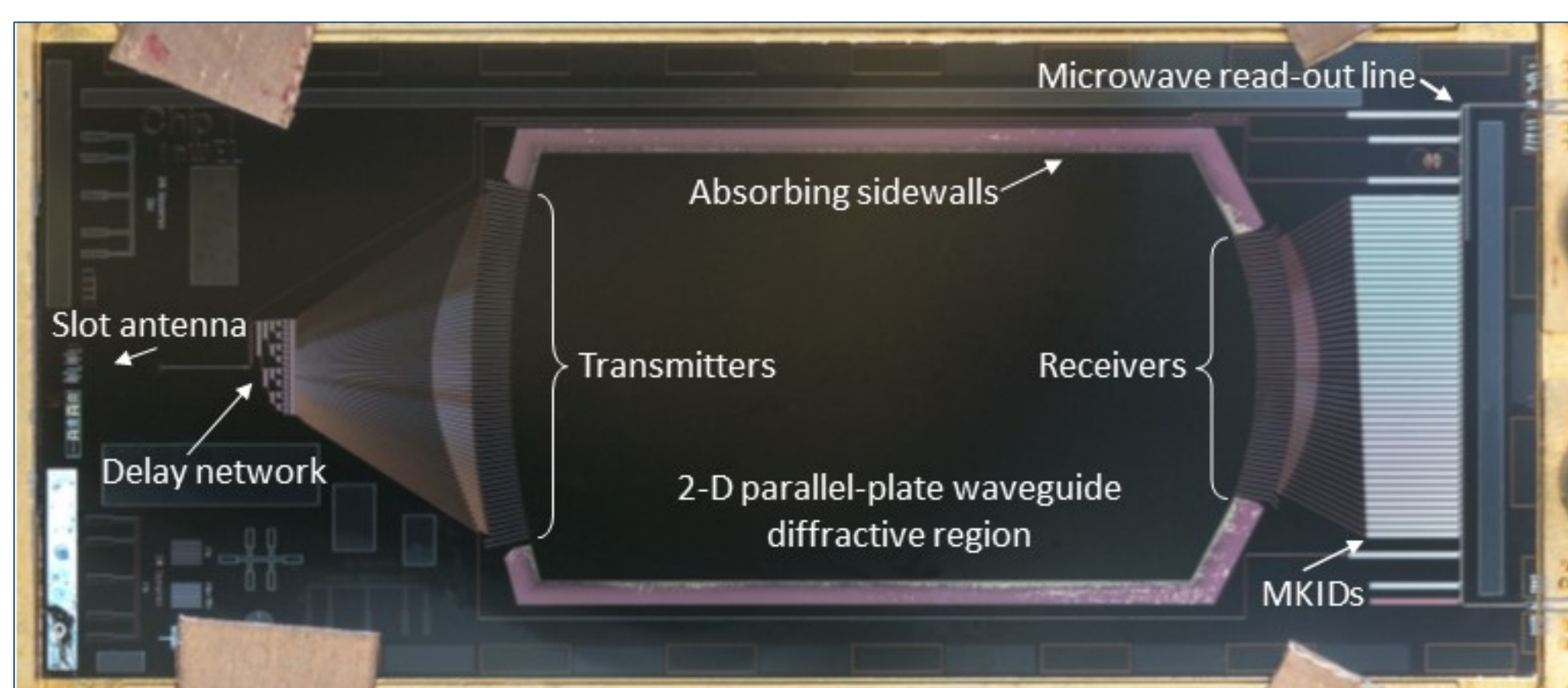


Figure 2. A test version with resolving power $R=64$ was designed [9], built and tested at NASA GSFC. The fabrication process developed [10] was employed to build several prototypes. The successful instrument optical performance tests have enabled us to demonstrate the μ -Spec technology [11].

Diffractive region design methodology

An **optimized geometry** is found for the transmitting antennas (red arc in Fig. 4) and the delay network extra lengths in silicon (not shown) through the minimization of the root-mean-square (RMS) phase error on the spectrometer focal plane (Fig. 5) [12]. The receiving antennas (green arc in Fig. 4) are laid to Nyquist-sample the combined power beam.

The design methodology was updated from [9,13] to account for the **dispersive effects** due to the superconductor's kinetic inductance (KI) in the silicon dielectric. The KI fraction is determined by the **niobium**, of which the transmission lines are made. The dispersive effects are quantified through the frequency dependence of the ratio of the phase velocity without KI to the phase velocity with KI following the Mattis-Bardeen theory [14].

1) INPUT PARAMETER SELECTION

Table 1: Spectrometer design parameters for the configuration selected to fit 2 spectrometers in a 15-cm-diameter silicon wafer.

Input parameter	Value
Min frequency	420 GHz
Max frequency	540 GHz
Resolving power	512
Grating design order	2
Diffractive region radius	1.35 cm
Stigmatic points' coordinates	See Fig. 4
Stigmatic points' frequencies	See Fig. 4
Silicon relative permittivity	11.55

2) DIFFRACTIVE REGION OPTIMIZATION

- 3a) Optimized design variables:
1) Transmitting antennas' x-y coordinates
2) Delay-network extra lengths in silicon
- 3b) RMS phase error minimization

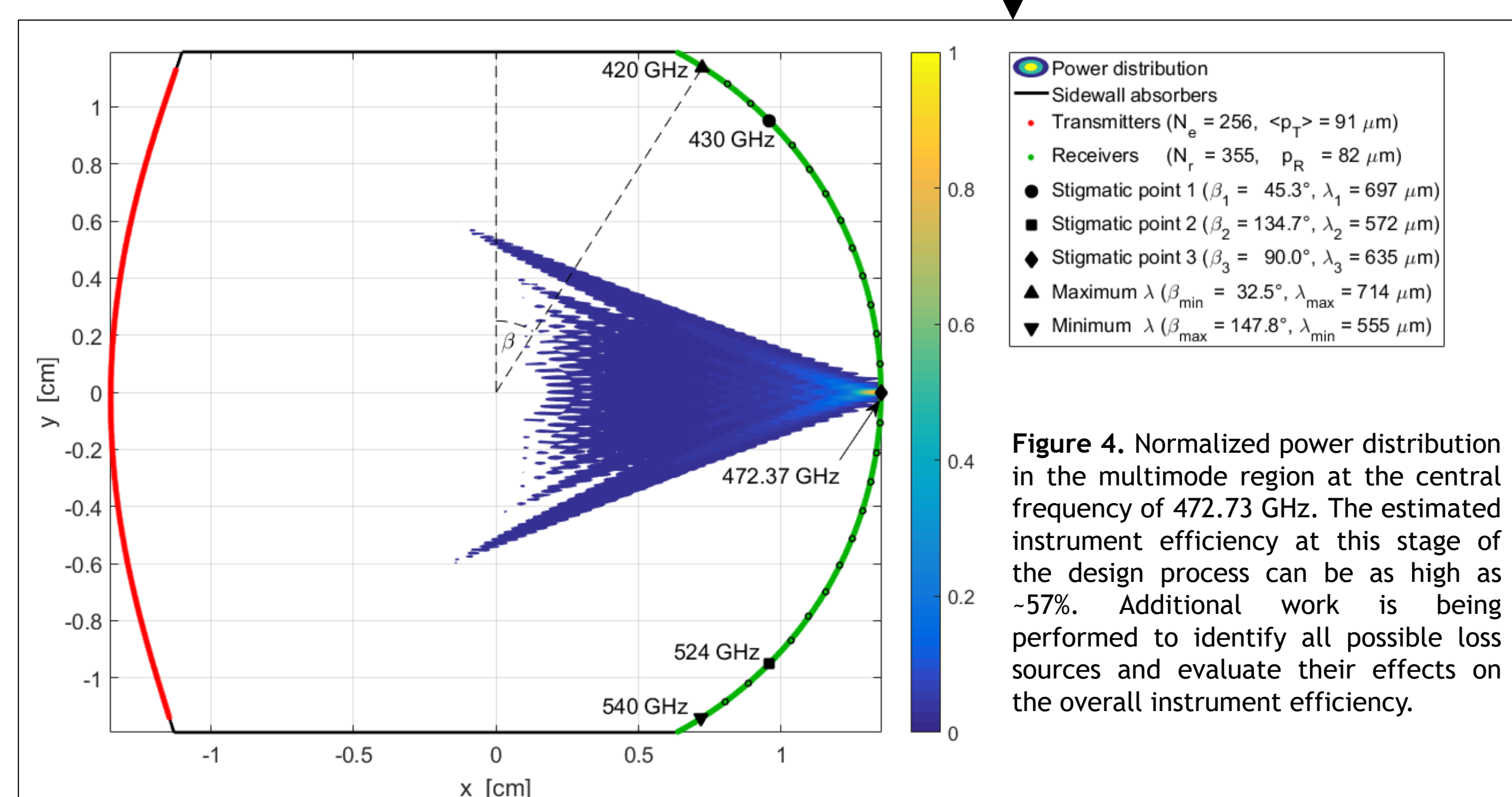


Figure 4. Normalized power distribution in the multimode region at the central frequency of 472.73 GHz. The estimated instrument efficiency at this stage of the design process can be as high as $\sim 57\%$. Additional work is being performed to identify all possible loss sources and evaluate their effects on the overall instrument efficiency.

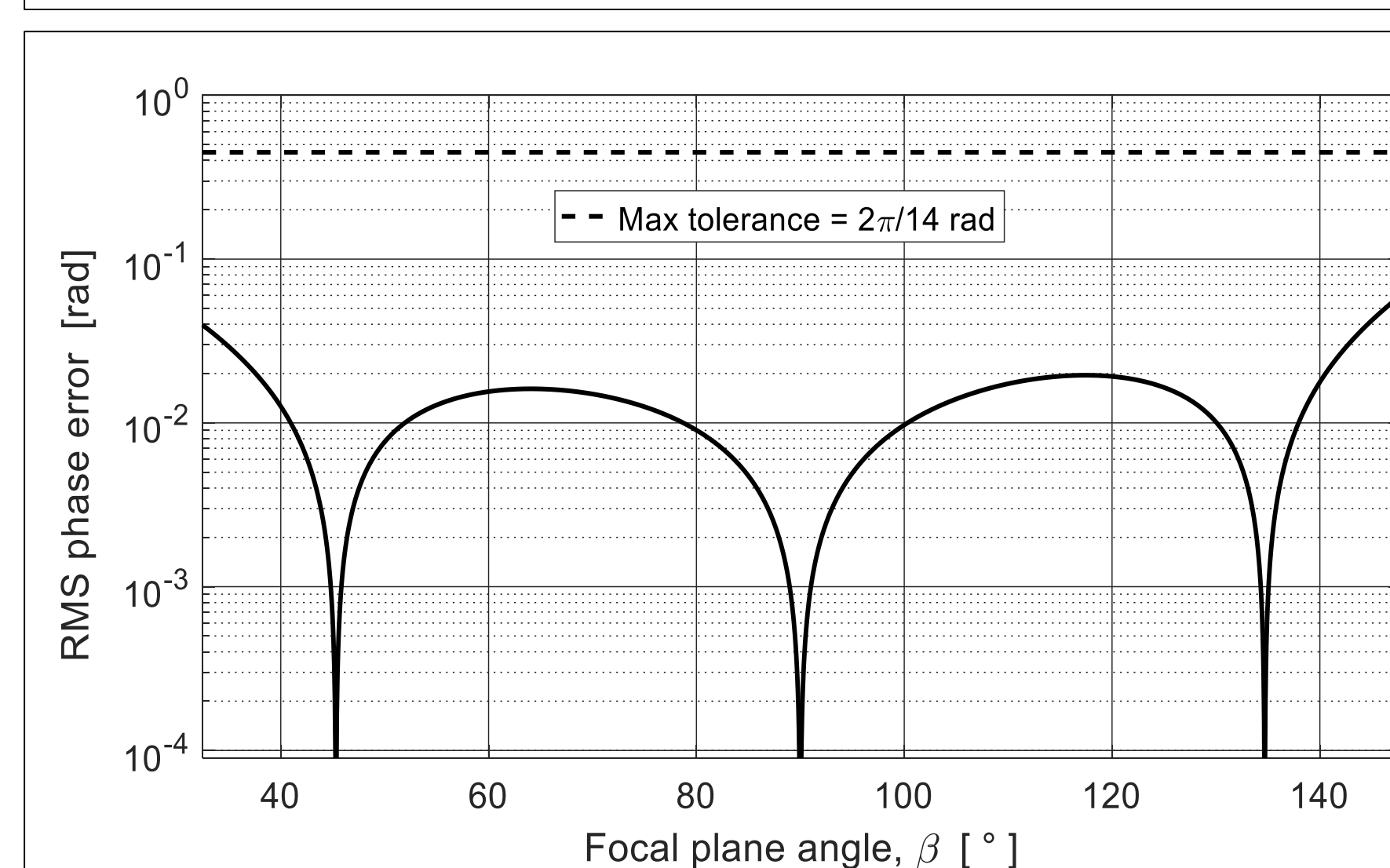


Figure 5. The positions and frequencies of the stigmatic points are selected so that the average pitch of the transmitting and receiving antenna arrays does not differ by more than approximately $\pm 10\%$. This ensures the two antenna arrays have the correct coupling. The RMS phase error vanishes at the three stigmatic points by construction and remains below the maximum allowed tolerance of $2\pi/14$ rad over the entire focal plane. This ensures **diffraction-limited performance** over the power combiner's design spectral range.

Delay network

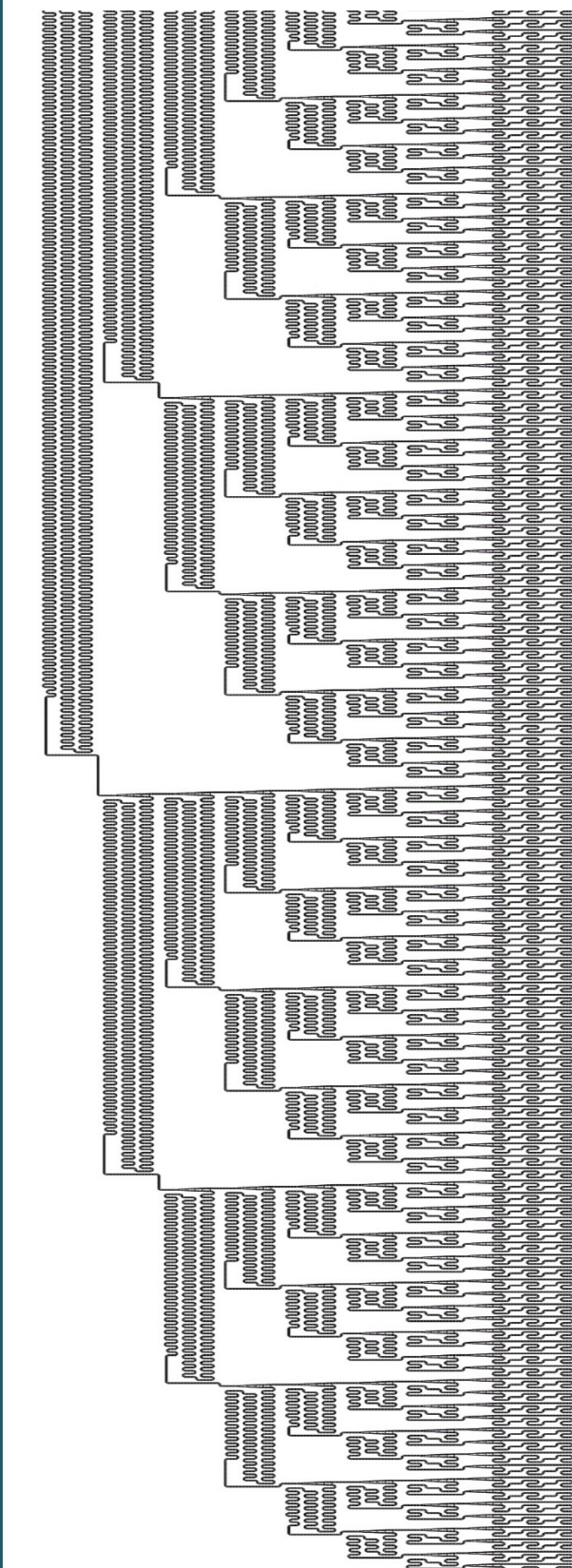
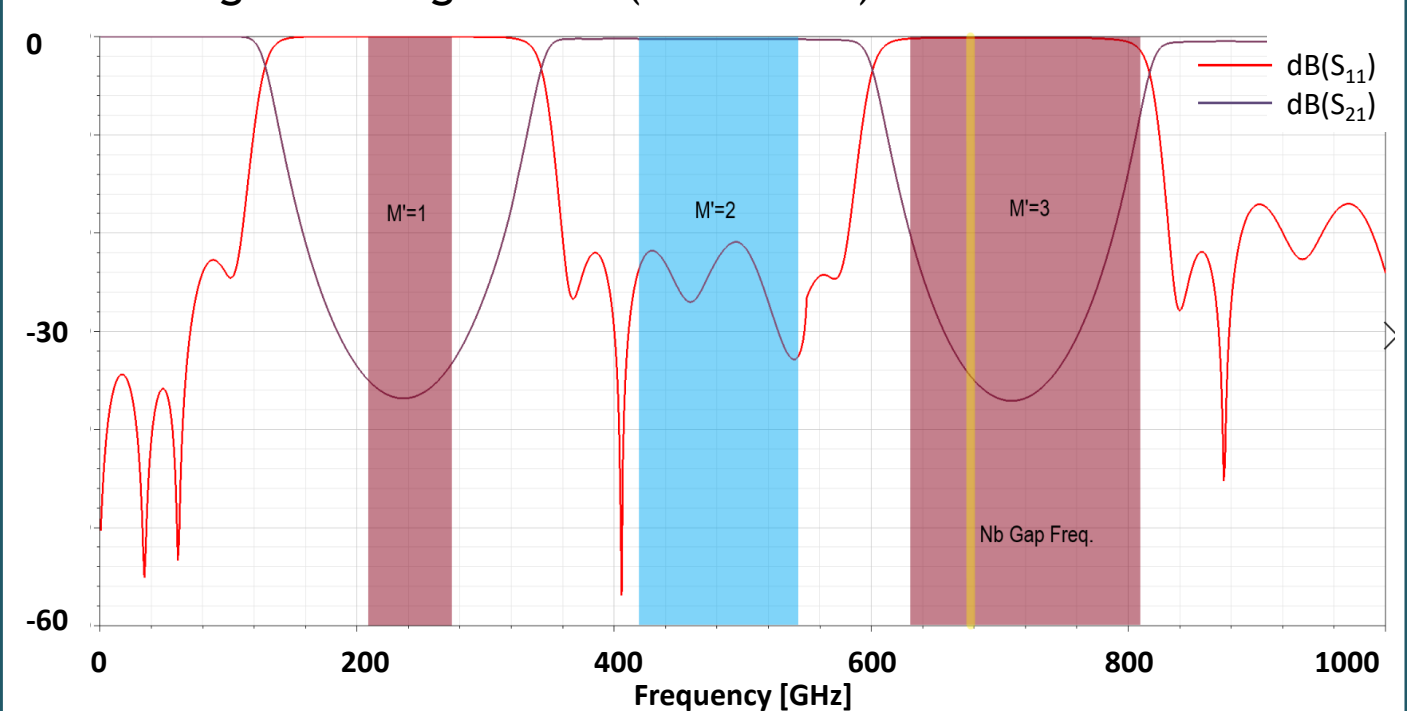


Figure 7. Layout of the μ -Spec delay network. The delay network consists of a binary splitting network of meandered microstrip transmission lines made of niobium (Nb) on a $0.45\text{-}\mu\text{m}$ -thick silicon (Si) dielectric substrate. This configuration reduces the required area by a factor of $\log_2(N)/(N-1) = 0.03$.

Each delay line's length is designed to generate the required phase retardation at the input to the planar diffractive region.

Given measured uniformities in the patterned Nb transmission lines and in the Si dielectric layer, our studies show we can achieve the phase control necessary to provide resolving powers ≥ 2300 .

Figure 8. Order-choosing filter passband. Before the light propagates through the delay network, an on-chip microstrip stepped-impedance bandpass filter enables selecting the design order (here $M'=2$).



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Acknowledgments

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